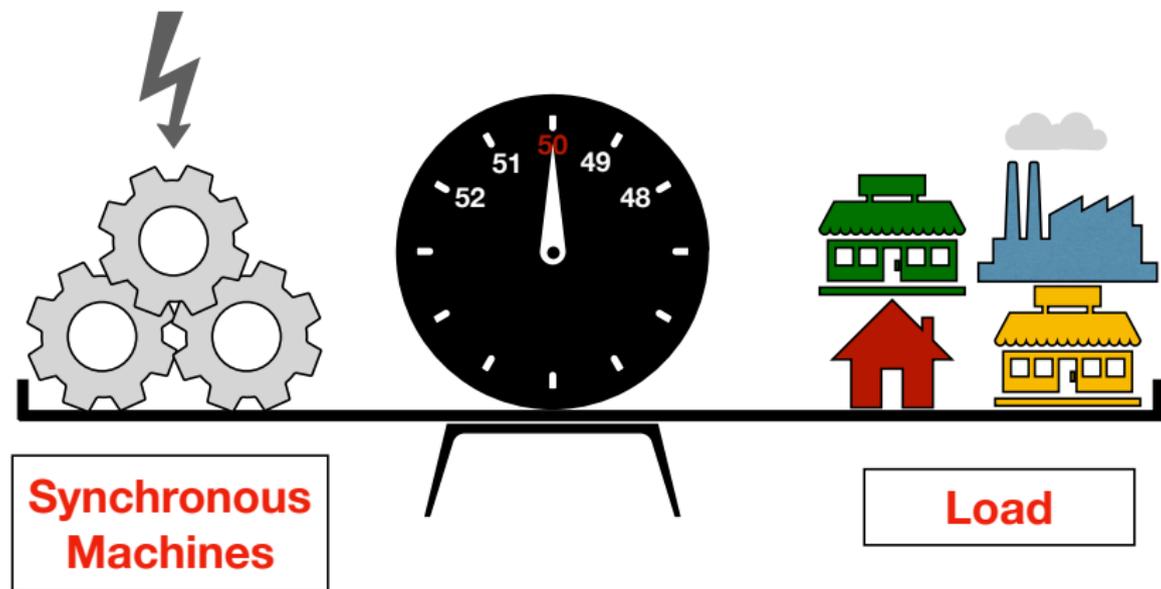




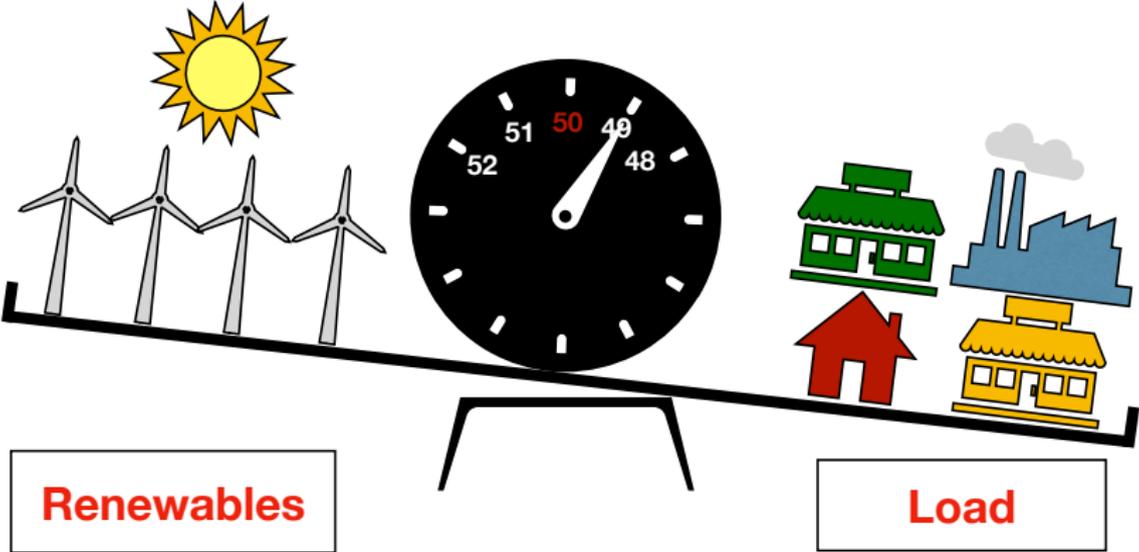
Parametric Local Stability of a Multi-Converter System

Taouba Jouini, Automatic Control Laboratory (IfA)

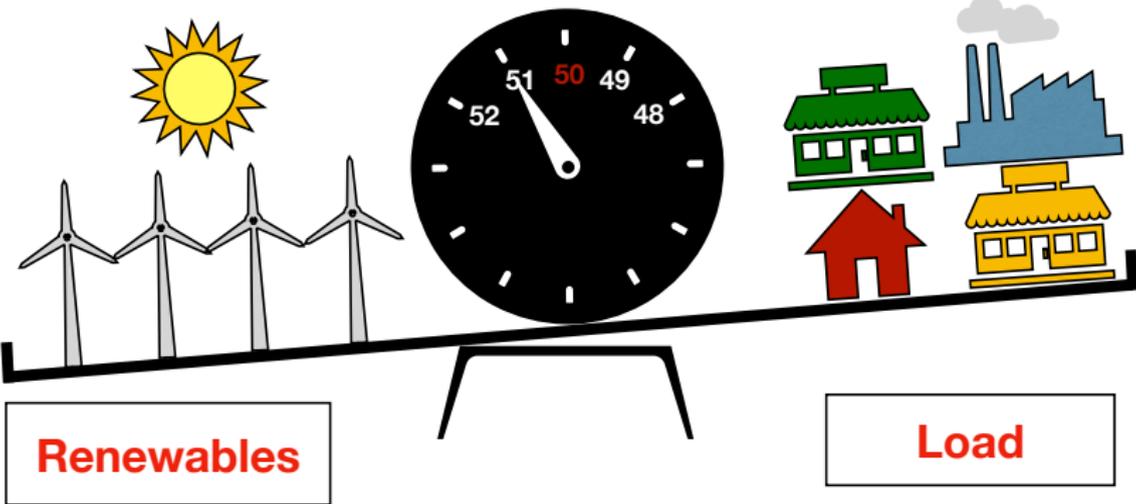
Motivation: classical operation of power system



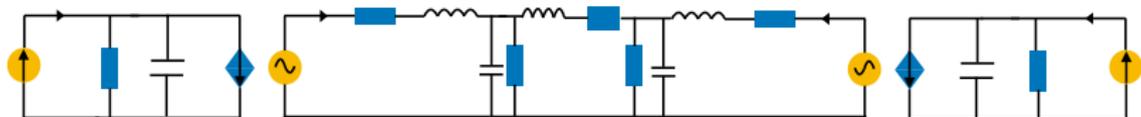
Motivation: changes in grid operation



Motivation: changes in grid operation



Modelling from first-order principle



Converter 1

RL Line

Converter 2

Balanced, three-phase, averaged DC/AC converter in ***dq*-frame** $\int_0^\tau \omega^* d\tau$

$$C_{dc} \dot{v}_{dc} = -G_{dc} v_{dc} - \frac{1}{2} \text{diag}(m^\top) i_{dq} + i_{dc}$$

$$L \dot{i}_{dq} = -\mathbf{Z}_R i_{dq} + \frac{1}{2} \text{diag}(m) v_{dc} - v_{dq}$$

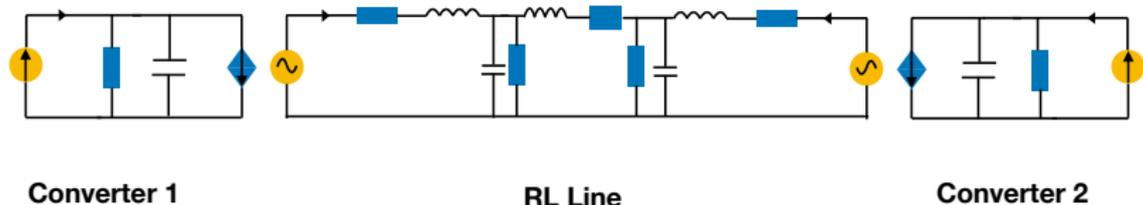
$$C \dot{v}_{dq} = -\mathbf{Z}_V v_{dq} + i_{dq} - \mathbf{B} i_{net,dq}$$

$$L_{net} \dot{i}_{net,dq} = -\mathbf{Z}_{net} i_{net,dq} + \mathbf{B}^\top v_{dq}$$

B: incidence matrix,

$\mathbf{Z}_R = R\mathbf{I} + \mathbf{J}\omega^*L$, $\mathbf{Z}_V = G\mathbf{I} + \mathbf{J}\omega^*C$, $\mathbf{Z}_{net} = R_{net}\mathbf{I} + \mathbf{J}\omega^*L_{net}$: impedance matrices, with $\mathbf{J} = \mathbf{I} \otimes \begin{bmatrix} 0 & -1 \\ 1 & 0 \end{bmatrix}$

Synchronous machine matching control



- ▶ synchronous machine *matching control*

$$\dot{\gamma} = \eta (v_{dc} - v_{dc}^*), \eta > 0,$$

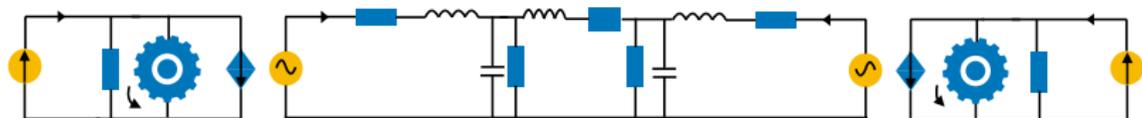
$$m = \mu \begin{bmatrix} -\sin(\gamma) \\ \cos(\gamma) \end{bmatrix}, 1 > \mu > 0$$

- ▶ *controllable* current source

$$i_{dc} = i_{dc}^* - K_p (v_{dc} - v_{dc}^*) - G_{dc} v_{dc}^*, K_p > 0$$

[Jouini et al.'16, Arghir et al.'18]

dq trafo with respect to desired frequency



Converter 1

RL Line

Converter 2

Closed-loop DC/AC converter in **dq-frame**

$$\dot{\gamma} = \eta (v_{dc} - v_{dc}^* \mathbf{1}_n)$$

$$C_{dc} \dot{v}_{dc} = -\hat{K}_p (v_{dc} - v_{dc}^* \mathbf{1}_n) - \frac{1}{2} \text{diag}(\mu) \mathbf{R}(\gamma)^\top i_{dq} + i_{dc}^*$$

$$L \dot{i}_{dq} = -\mathbf{Z}_R i_{dq} + \frac{1}{2} \text{diag}(\mu) \mathbf{R}(\gamma) v_{dc} - v_{dq}$$

$$C \dot{v}_{dq} = -\mathbf{Z}_V v_{dq} + i_{dq} - \mathbf{B} i_{net,dq}$$

$$L_{net} \dot{i}_{net,dq} = -\mathbf{Z}_{net} i_{net,dq} + \mathbf{B}^\top v_{dq}$$

where $\hat{K}_p = K_p + G_{dc}$, $K_p > 0$

Steady state manifold \mathcal{M} : Invariance under rotation angle θ_0

- ▶ steady-state **angles**

$$[\gamma] = \{\gamma \in \mathbf{T}_n \mid \omega^* t + \gamma(0) + \theta_0 \text{ span}\{\mathbf{1}_n\}\}$$

- ▶ steady-state **frequency** and **DC voltages**

$$[\omega] = \{\omega \in \mathbf{R}_{\geq 0}^n \mid \omega = \omega^*\}$$

$$[v_{dc}] = \{v_{dc} \in \mathbf{R}_{\geq 0}^n \mid v_{dc} = v_{dc}^*\}$$

- ▶ steady-state **AC quantities**

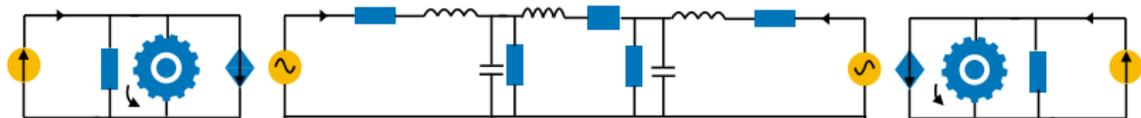
$$[i_{dq}] = \{i_{dq} \in \mathbf{R}^{2n} \mid i_{dq}^* + \text{span}\{\mathcal{R}(\theta_0)i_{dq}^*\}\}$$

$$[v_{dq}] = \{v_{dq} \in \mathbf{R}^{2n} \mid v_{dq}^* + \text{span}\{\mathcal{R}(\theta_0)v_{dq}^*\}\}$$

$$[i_{net,dq}] = \{i_{net,dq} \in \mathbf{R}^{2m} \mid i_{net,dq}^* + \text{span}\{\mathcal{R}(\theta_0)i_{net,dq}^*\}\}$$

$\mathcal{R}(\theta_0)$: rotating matrix with angle θ_0

Linearisation around a desired steady state x^*



Converter 1

RL Line

Converter 2

Linearised converter model at $x^* = [\gamma^* \ v_{dc}^* \ \mathbf{1}_n \ i_{dq}^* \ v_{dq}^* \ i_{net,dq}^*]^\top$

$$\dot{x} = \left[\begin{array}{cc|ccc} 0 & \eta \mathbf{I} & 0 & 0 & 0 \\ -\mathbf{A}_{21} & -C_{dc}^{-1} \hat{K}_p \mathbf{I} & -\mathbf{A}_{32}^\top & 0 & 0 \\ \hline \mathbf{A}_{31} & \mathbf{A}_{32} & -L^{-1} \mathbf{Z}_R & -L^{-1} \mathbf{I} & 0 \\ 0 & 0 & C^{-1} \mathbf{I} & -C^{-1} \mathbf{Z}_v & -C^{-1} \mathbf{B} \\ 0 & 0 & 0 & L_{net}^{-1} \mathbf{B}^\top & -L_{net}^{-1} \mathbf{Z}_{net} \end{array} \right] x$$

$$v(x^*) = \text{span} \{ [\mathbf{1}_n \ \mathbf{0} \ \mathbf{J} \ i_{dq}^* \ \mathbf{J} \ v_{dq}^* \ \mathbf{J} \ i_{net,dq}^*] \} \in \ker(A(x^*)), \quad v(x^*) = [v_1(x^*) \ v_2(x^*)]$$

Definition of a steady state tangent space $T_x^* \mathcal{M}$

- ▶ tangent **angles** subspace

$$[\gamma] = \{\gamma \in \mathbf{T}_n \mid \omega^* t + \gamma(0) + \text{span}\{\mathbf{1}_n\}\}$$

- ▶ tangent **frequency** and **DC voltages** subspace

$$[\omega] = \{\omega \in \mathbf{R}_{\geq 0}^n \mid \omega = \omega^*\}$$

$$[v_{dc}] = \{v_{dc} \in \mathbf{R}_{\geq 0}^n \mid v_{dc} = v_{dc}^*\}$$

- ▶ tangent **AC quantities** subspace

$$[i_{dq}] = \{i_{dq} \in \mathbf{R}^{2n} \mid i_{dq}^* + \text{span}\{\mathbf{J} i_{dq}^*\}\}$$

$$[v_{dq}] = \{v_{dq} \in \mathbf{R}^{2n} \mid v_{dq}^* + \text{span}\{\mathbf{J} v_{dq}^*\}\}$$

$$[i_{net,dq}] = \{i_{net,dq} \in \mathbf{R}^{2m} \mid i_{net,dq}^* + \text{span}\{\mathbf{J} i_{net,dq}^*\}\}$$

Separable Lyapunov analysis for systems with symmetries

A class of *partitioned* linear systems

$$\dot{x} = \left[\begin{array}{c|c} A_{11} & A_{12} \\ \hline A_{21} & A_{22} \end{array} \right] x \quad (5)$$

Three assumptions to go:

- ▶ the sub-block A_{11} is **Hurwitz**
- ▶ there **exists** a vector $v = [v_1 \ v_2]^\top : A \cdot \text{span}\{v\} = 0$
- ▶ ... (added in a later step)

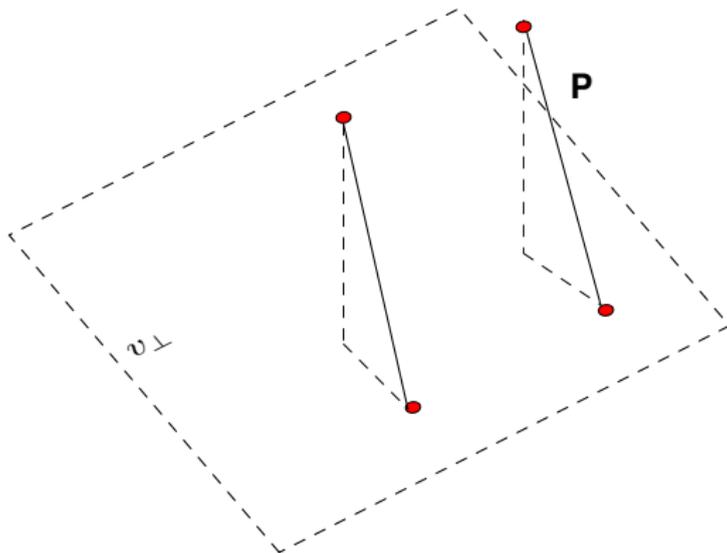
Derive **fully decentralized** conditions so that $\text{span}\{v\}$ is **asymptotically stable** \Leftrightarrow the system matrix has **all its eigenvalues** in the **left-half** plane except for **one** at **zero**

Construction of the Lyapunov function

- ▶ define LF as an *oblique projector* into v_{\perp} in \mathbf{P} -columns direction

$$\mathcal{V}(x) = x^{\top} \left(\mathbf{P} - \frac{\mathbf{P}v v^{\top} \mathbf{P}}{v^{\top} \mathbf{P} v} \right) x : A \cdot \text{span}\{v\} = 0,$$

\mathbf{P} is a *positive definite* matrix.



Separable Lyapunov function

- ▶ fix the **structure** of the matrix $Q(\mathbf{P})$ so that $\mathbf{P}A + A^\top \mathbf{P} = -Q(\mathbf{P})$

$$Q(\mathbf{P}) = \left[\begin{array}{c|c} \mathbf{Q}_1 & H(\mathbf{P})^\top \\ \hline H(\mathbf{P}) & H(\mathbf{P})\mathbf{Q}_1^{-1}H^\top(\mathbf{P}) + \mathbf{Q}_2 \end{array} \right], H(\mathbf{P}) = A_{12}^\top \mathbf{P}_1 + \mathbf{P}_2 A_{21}$$

\mathbf{Q}_1 is **positive definite**, \mathbf{Q}_2 is **positive semi-definite** with respect to v_2

- ▶ choose block diagonal matrix \mathbf{P}

$$\mathbf{P} = \left[\begin{array}{c|c} \mathbf{P}_1 & 0 \\ \hline 0 & \mathbf{P}_2 \end{array} \right], \mathbf{P}_1 = \mathbf{P}_1^\top, \mathbf{P}_2 = \mathbf{P}_2^\top$$

- ▶ $\mathbf{P}A + A^\top \mathbf{P} = -Q(\mathbf{P})$ Algebraic Riccati Equation
- ▶ **Third** assumption: define $\mathbf{F} = A_{22} + A_{21}\mathbf{Q}_1^{-1}P_1A_{12}$
 - **necessary** and **sufficient** condition: $(\mathbf{F}, A_{21}\mathbf{Q}_1^{-1/2})$ is stabilizable and (\mathbf{F}, D) is detectable, where $DD^\top = A_{12}^\top P_1 \mathbf{Q}_1^{-1} P_1 A_{12} + \mathbf{Q}_2$
 - \mathbf{F} is a **Hurwitz** matrix (stricter condition)

Main result and contextualization

If **three** assumptions are satisfied, then $\text{span}\{v\}$ is **asymptotically stable**

The matrix \mathbf{F} is Hurwitz:

- ▶ **small-gain** theorem for A_{22}
Hurwitz
- ▶ **passive** system: LE, KYP
Lemma

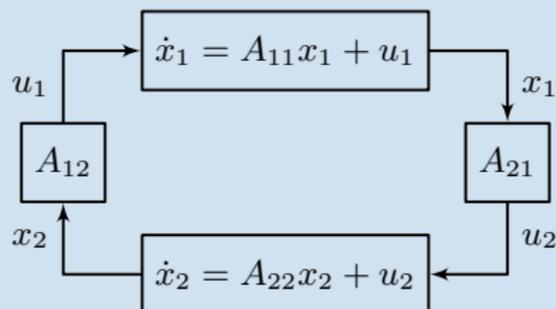
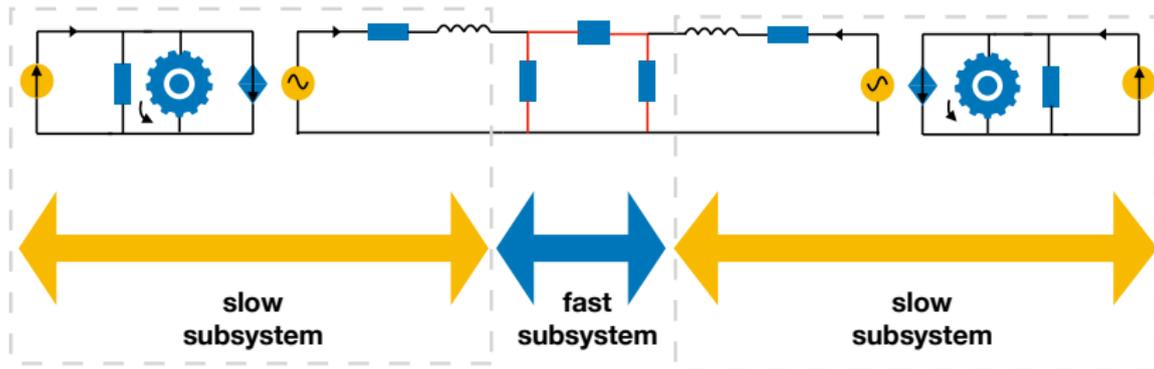


Figure: Interconnected closed-loop system

ECC Paper: application to reduced model



- ▶ application of *Tikhonov's* Theorem
- ▶ *slow* subsystem: $\{\gamma, v_{dc}, i_{dq}\}$ and *fast* subsystem: $\{v_{dq}, i_{net,dq}\}$
- ▶ fast subsystem is *exponentially stable* relative to $[v_{dq}^* \ i_{net,dq}^*]^\top$
- ▶ *sufficient* and *fully decentralized* condition: reactive power support and resistive damping
- ▶ arXiv paper: *full-order* model analysis [Jouini et al.'19]

Application: Local stability of multi DC/AC converter

Algebraic synchronization condition at the k -th converter:

$$16 Q_{x,k}^* > \frac{\mu_k^2 v_{dc}^{*2}}{R} \quad (6)$$

where $Q_{x,k}^* = \frac{1}{2} v_{dc}^* \mu_k J r(\gamma_k^*)^\top i_{dq}^*$, $\mathbf{Q}_1 = \mathbf{I}$, $\mathbf{Q}_2 = \mathbf{I} - \frac{v_2(x^*) v_2(x^*)^\top}{v_2(x^*)^\top v_2(x^*)}$

- ▶ **sufficient** condition (6) is **fully decentralized**
- ▶ if (6) is **true**, then $\text{span}\{v(x^*)\}$ is **locally** asymptotically stable
- ▶ **sufficient** reactive power support, **resistive damping**
- ▶ reinforced by **virtual** impedance control:

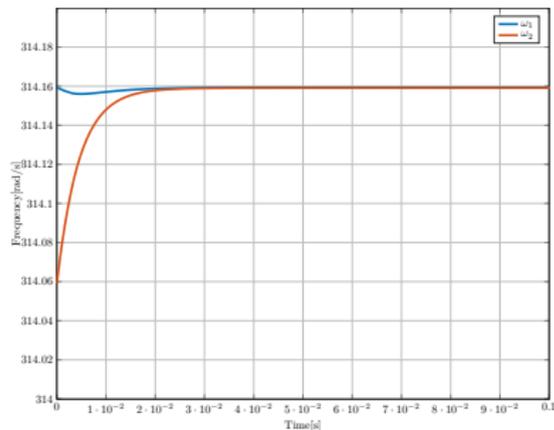
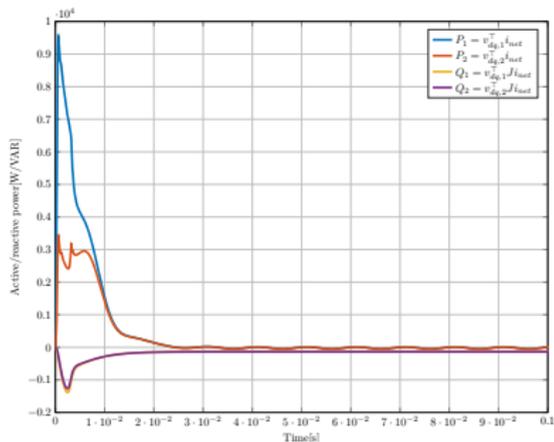
$$m' = m + 2k_m i_{dq} / v_{dc}, i'_{dc} = i_{dc} + k_m i_{dq}^\top i_{dq} / v_{dc}, k_m > 0$$

- ▶ aligned with **practitioners** insights [Wang et al.'15, Barbanov et al.'16]

Simulations: Eye candy

- ▶ identical DC/AC converters initialized differently
 $v_{dc,1}(0) \neq v_{dc,2}(0)$
- ▶ same input pair (i_{dc}^*, μ^*)
- ▶ stability condition satisfied for $k_m > 47.84$

	Converter 1	Converter 2	RL Line
i_{dc}^*	$1.9 \cdot 10^3$	$1.9 \cdot 10^3$	—
v_{dc}^*	10^3	10^3	—
C_{dc}	10^{-3}	10^{-3}	—
G_{dc}	10^{-5}	10^{-5}	—
η	10^{-4}	10^{-4}	—
R	0.1	0.1	—
L	$5 \cdot 10^{-4}$	$5 \cdot 10^{-4}$	—
C	10^{-5}	10^{-5}	—
μ^*	0.33	0.33	—
G	0.05	0.05	—
K_p	0.2	0.2	—
R_{net}	—	—	2
L_{net}	—	—	10^{-4}



Reminder: generic analysis challenges

- C1. assumptions on *quasi-stationary* steady state
- C2. transmission lines with *non-zero* transfer conductance
- C3. interaction of grid units with *line dynamics*
- C4. *dq-frame* in multi-machine case study
- C5. need for *fully decentralized* stability conditions

Our remedy in this work

- R1. modelling from *first-order* principles
- R2. stability analysis of *reduced-order* DC/AC converter model
- R3. extensions to *full-order* DC/AC converter model
- R4. dq-transformation with respect to *steady state frequency*
- R5. scalable LF: *parametrised* Lyapunov/Ricatti equation

Future venues

- ▶ study of conservativeness of derived stability condition
- ▶ extensive simulations of multi-converter system

Thank you for your attention